EE 317
Spring 2001
Midterm Exam #2

Name:	KEY		

ID Number:

This exam consists of 6 pages.

The problems are worth a total of 100 points. The number of points you can earn on each problem is given in [brackets] next to the problem number.

For all problems, clearly show all work.

Where appropriate, enclose final answers in boxes, and include units in all answers.

Calculators are permitted.

You are permitted to consult your textbook but not any other notes of any kind.

1. [2 pts/characteristic] A certain signal x(t) has the Fourier series in trigonometric form given by:

$$x(t) = \sum_{n=1}^{\infty} \frac{(-1)^n}{4\pi n} \sin(40\pi nt)$$

Circle the appropriate characteristics of the signal based upon what you know about its Fourier series: conflict exam: even

even function

odd function) - sine function is odd; sum of sines is odd

periodic

aperiodic

- any signal that has a Fourier series representation is periodic

energy signal

(power signal) - bounded periodic signals are power signals ( on energy, finite power)

continuous-frequency spectrum

discrete-frequency - frequency components only at spectrum multiples of ωo (ωo = 40π in this case)

real-valued

- sine is real-valued; Fourier coef & complex-valued (cn) in this case are real-valued (no

j's anywhere!)

2. [0, 4 points] What is the fundamental period of the signal in Prob. 1?

- a)  $1/4\pi$  sec
- b)  $1/40\pi \sec$
- c) 25 msec

- d))50 msec
- e)  $40\pi$  sec
- f) not enough information to solve

$$\omega_o = 40\pi$$
  $T_o = \frac{2\pi}{\omega_o} = \frac{2\pi}{40\pi} = \frac{1}{20} = 0.05 \text{ sec}$   
= 50 m/sec

conflict exam: To = 25ms

For Problems 3-6, refer to the following exponential Fourier series of the signal f(t). Note that the summation has a finite number of terms.

$$f(t) = 0.3 + \sum_{\substack{n = -2 \\ n \neq 0}}^{2} \frac{j(-1)^{n}}{n|n|} e^{j6280nt} \qquad \omega_{o} = 6280 \text{ rad/s}$$

$$f_{o} = \frac{\omega_{o}}{2\pi} = \frac{6280}{2\pi} = 999 \text{ Hz}$$

[0, 2, 4 points] What is the total normalized average power of the signal?

$$P = \frac{c^2 + 2|c_1|^2 + 2|c_2|^2}{|c_n|^2 + 2\left(\frac{1}{1^2}\right)^2 + 2\left(\frac{1}{2^2}\right)^2}$$

$$= \frac{0.3^2 + 2\left(\frac{1}{1^2}\right)^2 + 2\left(\frac{1}{2^2}\right)^2}{|c_n|^2 + 2\left(\frac{1}{2^2}\right)^2}$$

$$= \frac{1}{|c_n|} = \frac{1}{|c_n|} = \frac{1}{|c_n|}$$

$$= \frac{1}{|c_n|} = \frac{1}{|c_n|}$$

4. [0, 4 points] What percentage of the total power in the signal lies below 1200 Hz?

a) 0% b) 4.1% c) 60% d) 90% (e) 94% f) 100% signal has frequency % power = 
$$(100\%) \frac{0.3^2 + 2(1)^2}{2.215} = 94\%$$
 components at  $f = (0)(999), (1)(999),$  and  $(2)(999)$  [0, 4 points] What is the time-average value of the signal?  $= 0, 999, 1998 H_3$ 

5.

a) 0 b) 0.09 c) 0.15 d) 0.3 e) 0.188 f) not enough info 
$$F(t)|_{a \vee g} = C_o = o.3$$

$$conflict exam: 0.4$$

6. [2 pts each] What is the phase (in radians) of each coefficient in the Fourier series?

$$\angle C_{-2} = \frac{-\pi/z}{2}$$

$$\angle C_{-1} = \frac{3(-1)^{-2}}{-2(-1)} = \frac{3(1)}{-2(2)} = -3\frac{1}{4} = \frac{1}{4} \angle -\pi/z$$

$$\angle C_{-1} = \frac{3(-1)^{-1}/-1}{-1/-1} = -3/-1 = 3 = 1 \angle \pi/z$$

$$\angle C_{0} = \frac{0}{2}$$

$$\angle C_{0} = \frac{0}{2}$$

$$\angle C_{0} = \frac{0.3 \text{ (by in spection)}}{2} = 0.3 \angle 0 \text{ (real)}$$

$$\angle C_{1} = \frac{3(-1)^{2}/-1}{2} = -3 = 1 \angle -\pi/z$$

$$\angle C_{2} = \frac{3(-1)^{2}}{2|2|} = \frac{1}{4} = \frac{1}{4} \angle \pi/z$$

ronflict exam: same answers 7. [0-16 points] Find the Fourier transform of the following signal. You may use the direct integration approach, or you may use Tables 5.1 and 5.2 in your text to minimize the work required. (For full credit, you must show **all** of your work.)

$$f(t) = e^{-at} \cos(\omega_0 t) u(t)$$
Let  $f_1(t) = e^{-at} u(t)$   $\Longrightarrow$   $F_1(\omega) = \frac{1}{a+j\omega}$   $\Longrightarrow$  Table 5.2

Let  $f_2(t) = \cos(\omega_0 t)$   $\Longrightarrow$   $F_2(\omega) = \pi \left[ \mathcal{E} \left( \omega - \omega_0 \right) + \mathcal{E} \left( \omega + \omega_0 \right) \right]$ 

$$f(t) = f_1(t) f_2(t) \Longrightarrow F(\omega) = F_1(\omega) \times F_2(\omega) \quad (convolution prop)$$

$$Table 5.1$$

$$\Longrightarrow F(\omega) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F_1(x) F_2(\omega - x) dx \quad (x = dummy Variable)$$

$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{1}{a+j\lambda} \sqrt{\left[ \mathcal{E} \left( \omega - x - \omega_0 \right) + \mathcal{E} \left( \omega - x + \omega_0 \right) \right] dx}$$

$$= \frac{1}{2} \int_{-\infty}^{\infty} \frac{1}{a+j\lambda} \sqrt{\left[ \mathcal{E} \left( \omega - x - \omega_0 \right) + \mathcal{E} \left( \omega - x + \omega_0 \right) \right] dx}{\left[ \alpha + \frac{1}{2} \left( \omega - \omega_0 \right) - x \right] dx + \frac{1}{2} \int_{-\infty}^{\infty} \frac{\mathcal{E} \left[ \left( \omega + \omega_0 \right) - x \right] dx}{\left[ \alpha + \frac{1}{2} \left( \omega - \omega_0 \right) \right] \left[ \alpha + \frac{1}{2} \left( \omega + \omega_0 \right) \right]}$$

$$= \frac{1}{2} \frac{1}{a+j\lambda} \frac{1}{(\omega + \omega_0)} + \frac{1}{2} \frac{1}{a+j\lambda} \frac{1}{(\omega + \omega_0)}$$

$$= \frac{1}{2} \frac{1}{(a+j\lambda)} \frac{1}{(\omega + \omega_0)} \frac{1}{(a+j\lambda)} \frac{1}{(\omega + \omega_0)}$$

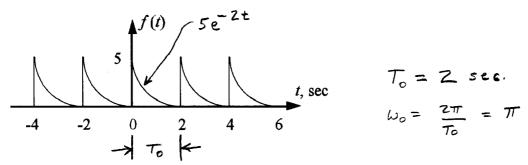
$$= \frac{1}{2} \frac{1}{(a+j\lambda)} \frac{1}{(\omega + \omega_0)} \frac{1}{(a+j\lambda)} \frac{1}{(\omega + \omega_0)}$$

$$= \frac{1}{2} \frac{1}{(a+j\omega)^2 + \omega_0^2} \frac{1}{(a+j\omega)^2 + (a+j\omega)} \frac{1}{(a+j\omega)^2 + (a+j\omega)^2 + (a+j\omega)$$

conflict exam:

$$F(\omega) = \frac{2 + j \omega}{(2 + j \omega)^2 + (5\pi)^2}$$

8. [0-16 points] Find the Fourier series coefficients for the following periodic signal,



which can be expressed mathematically in the time domain as

$$f(t) = \sum_{n=-\infty}^{\infty} 5e^{-2(t-2n)} \operatorname{rect}\left(\frac{t-1-2n}{2}\right) \qquad eqn (4.23): c_{k} = \frac{1}{T_{o}} \int_{c_{To}}^{\infty} f(t) e^{\frac{1}{2}k\omega_{o}t} dt$$

$$= \frac{1}{2} \int_{0}^{2} 5e^{-2t} e^{-\frac{1}{2}k\pi t} dt$$

$$= \frac{5}{2} \int_{0}^{2} e^{-(2+\frac{1}{2}k\pi)t} dt$$

$$= \frac{5}{2} \frac{-1}{2+\frac{1}{2}k\pi} e^{-(2+\frac{1}{2}k\pi)t} dt$$

$$= \frac{5}{2} \frac{-1}{2+\frac{1}{2}k\pi} \left[ e^{-(2+\frac{1}{2}k\pi)(2)} - 1 \right]$$

$$= \frac{-5}{4} \frac{1}{4} e^{-\frac{1}{2}k2\pi} dt$$

$$= \frac{5}{4} \left( e^{-\frac{1}{2}(2)} - 1 \right)$$

$$= \frac{-5}{4} \left( e^{-\frac{1}{2}(2\pi)} - 1 \right)$$

$$= \frac{5}{4} \frac{1}{4} e^{-\frac{1}{2}k2\pi} dt$$

$$= \frac{5}{4} \left( e^{-\frac{1}{2}(2\pi)} - 1 \right)$$

$$= \frac{5}{4} e^{-\frac{1}{2}k2\pi} dt$$

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$$= \frac{5}{4} e^{-\frac{1}{2}k2\pi} dt$$

$$= \frac{5}{4} \left( e^{-\frac{1}{2}(2\pi)} - 1 \right)$$

$$= \frac{5}{4} e^{-\frac{1}{2}k2\pi} dt$$

$$= \frac{1}{2} \int_{0}^{2} 5e^{-\frac{1}{2}t} dt$$

$$= \frac{5}{2} \left( e^{-\frac{1}{2}(2\pi)} - 1 \right)$$

$$= \frac{5}{4} \left( e^{-\frac{1}{2}(2\pi)} - 1 \right)$$

$$= \frac{5}{4} e^{-\frac{1}{2}k2\pi} dt$$

$$= \frac{1}{2} \int_{0}^{2} 5e^{-\frac{1}{2}t} dt$$

$$= \frac{5}{2} \left( e^{-\frac{1}{2}(2\pi)} - 1 \right)$$

$$= \frac{5}{4} \left( e^{-\frac{1}{2}(2\pi)} - 1 \right)$$

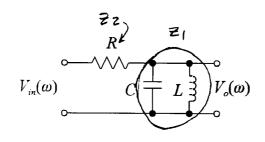
$$= \frac{5}{4} e^{-\frac{1}{2}k2\pi} dt$$

$$= \frac{1}{2} \int_{0}^{2} e^{-\frac{1}{2}k\pi} dt$$

$$= \frac{1}{2} \int_{0}^{2} e^{-\frac{1}{2$$

conflict exam:

$$C_k = \frac{6}{4 + 1' k \pi} \left( 1 - e^{-8} \right) \quad \begin{array}{c} 4 \\ C_0 = \frac{6}{4} \left( 1 - e^{-8} \right) = 1.50 \end{array}$$



For Problems 9-12, refer to the *RLC* circuit to the left.

$$Z_1 = \frac{1}{2mc} || 3mL = \frac{1}{2mc} (\frac{1}{2mc}) (\frac{1}{2mL})$$

$$Z_2 = R$$
 =  $\frac{L/c}{\sqrt{(\frac{L}{L} + \omega L)}}$ 

[12 points] Derive the transfer function  $H(\omega)$  for the circuit.

voltage-divider rule:

$$V_{0} = \frac{Z_{1}}{Z_{1}+Z_{2}} \text{ Vin } \Rightarrow H(\omega) = \frac{V_{0}}{Vin} = \frac{\frac{U_{C}}{\sqrt{3(\frac{1}{\omega_{C}}+\omega_{L})}}}{R + \frac{U_{C}}{\sqrt{3(\frac{1}{\omega_{C}}+\omega_{L})}}} = \frac{Z_{1}}{Z_{2}+Z_{1}}$$

Multiply numerator and denom. by 1/2:

$$H(\omega) = \frac{1}{1 + \frac{2}{2} \frac{1}{2}} = \frac{1}{1 + \frac{3(\frac{1}{2} \frac{1}{2} + \omega L)R}{\frac{1}{2}}} = \frac{1}{1 + \frac{3(\frac{1}{2} \frac{1}{2} + \omega RC)}{\frac{1}{2}}}$$

$$H(\omega) = \frac{1}{1 + j \left(\omega RC - \frac{R}{\omega L}\right)}$$

10. [0, 4, 8 points] At what frequency does the peak power transfer occur? The power transfer characteristic is given by  $|H(\omega)|^2$ .

$$|H(\omega)|^2 = H(\omega)H^*(\omega) = \frac{1}{|+j(\omega RC - \frac{R}{\omega L})} \cdot \frac{1}{|-j(\omega RC - \frac{R}{\omega L})}$$

$$= \frac{1}{1 + \left(\omega RC - \frac{R}{\omega L}\right)^2}$$

this quantity is alway positive (because of power of 2)

IHIZ has peak value at frequency that satisfies:

$$\omega RC - \frac{R}{\omega L} = 0 \implies \omega RC = \frac{R}{\omega L} \implies \omega C = \frac{1}{\omega L} \implies \omega^2 = \frac{1}{LC}$$

$$5 \implies \omega \rho = \frac{1}{\sqrt{LC}}$$

[0, 4, 8 points] What is/are the cut-off (half-power) frequency/frequencies of the circuit?

Half-power frequencies occur when 
$$|H(w)|^2 = \frac{1}{2} |H_{max}|^2$$
  
In this case  $|H_{max}|^2 = |H(w_{pk})|^2 = 1$ , so
$$|H(w_c)|^2 = \frac{1}{2} \quad \text{where } w_c = \text{cut-off frequency (ies)}$$

$$\Rightarrow |H(\omega_c)|^2 = \frac{1}{2} = \frac{1}{1 + (\omega_c RC - \frac{R}{\omega_c L})^2}$$

$$\Rightarrow (\omega_c RC - \frac{R}{\omega_c L})^2 = 1 \Rightarrow \omega_c RC - \frac{R}{\omega_c L} = \pm 1 \leftarrow Mult. \text{ by } \omega_c L$$

$$\omega_c^2 RLC - R = \pm \omega_c L$$

$$RLC \omega_c^2 \mp \omega_c L - R = 0 \leftarrow \text{div. by } RLC$$

$$\omega_{c}^{2} + \frac{1}{Rc} \omega_{c} - \frac{1}{Lc} = 0$$

$$\omega_{c} = \pm \frac{1}{2Rc} \pm \frac{1}{2} \sqrt{\left(\frac{1}{Rc}\right)^{2} + 4\left(\frac{1}{Lc}\right)^{2}}$$

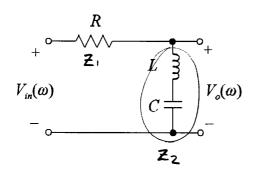
$$= \pm \frac{1}{2Rc} \pm \frac{1}{2} \sqrt{\left(\frac{1}{Rc}\right)^{2} \left[1 + \frac{4R^{2}c}{L}\right]^{2}}$$

quadratic formula:  

$$ax^2 + bx + c = 0$$
  
 $x = \frac{-b}{2a} \pm \frac{1}{2a} \sqrt{b^2 - 4ac}$ 

$$\left[\omega_{c} = \frac{1}{2RC} \left[ \frac{\pm 1 \pm \sqrt{1 + \frac{4R^{2}c}{L}}}{L} \right] \right] \leftarrow \text{far enough for full credit}$$

- 12. [0, 4 points] What ideal filter type most closely approximates the frequency response of the circuit?
  - a) lowpass
- b) highpass
- (c) bandpass
- d) bandstop
- 6  $|H(\omega)|^2$  rolls off (decreases in value) on either side of  $\omega = \omega \rho k$ .



For Problems 9-12, refer to the *RLC* circuit to the left.

$$Z_2 = \frac{1}{3wc} + 3wL$$

$$Z_1 = R$$

9. [12 points] Derive the transfer function  $H(\omega)$  for the circuit.

$$H(\omega) = \frac{Z_2}{Z_1 + Z_2} = \frac{\frac{1}{3\omega c} + 3\omega L}{R + \frac{1}{3\omega c} + 3\omega L} = \frac{1 - \omega^2 Lc}{1 - \omega^2 Lc + 3\omega Rc} = H(\omega)$$

10. [0, 4, 8 points] At what frequency does the **minimum** power transfer occur? The power transfer characteristic is given by  $|H(\omega)|^2$ .

$$|H(\omega)|^{2} = H(\omega) H(\omega) = \frac{1 - \omega^{2}LC}{1 - \omega^{2}LC + j\omega RC} \cdot \frac{1 - \omega^{2}LC}{1 - \omega^{2}LC - j\omega RC}$$

$$= \frac{(1 - \omega^{2}LC)^{2}}{(1 - \omega^{2}LC)^{2} + (\omega RC)^{2}} \cdot \frac{1}{\omega^{2}} = \frac{\left(\frac{1}{\omega} - \omega LC\right)^{2}}{\left(\frac{1}{\omega} - \omega LC\right)^{2} + (RC)^{2}}$$

$$|H(\omega)|^{2} = 0 \quad \text{when} \quad \left(\frac{1}{\omega} - \omega LC\right)^{2} = 0 \quad \Rightarrow \quad \omega_{min} = \frac{1}{\sqrt{LC}}$$

11. [0, 4, 8 points] What is/are the cut-off (half-power) frequency/frequencies of the circuit? Hint: Peak power transfer occurs at the frequency  $\omega = 0$ .

$$|H(\omega)|^2 = \frac{\left(\frac{1}{\omega} - \omega Lc\right)^2}{\left(\frac{1}{\omega} - \omega Lc\right)^2 + \left(Rc\right)^2} = \frac{1}{\left(\frac{1}{\omega} - \omega Lc\right)^2}$$

cut-off frequencies are those at which  $|H|^2 = \frac{1}{2} |H|_{max}^2$ 

$$|H|_{\text{max}}^2 = |H(0)|^2 = \frac{1}{1 + \frac{(RC)^2}{(\frac{1}{2} \cdot OLC)^2}} = 1$$

$$\Rightarrow |H(\omega_c)|^2 = \frac{1}{2} = \frac{1}{1 + \frac{(RC)^2}{\left(\frac{1}{\omega_c} - \omega_c LC\right)^2}} \Rightarrow \frac{(RC)^2}{\left(\frac{1}{\omega_c} - \omega_c LC\right)^2} = 1$$

$$(RC)^{2} = \left(\frac{1}{\omega_{c}} - \omega_{c} LC\right)^{2} \rightarrow \pm RC = \frac{1}{\omega_{c}} - \omega_{c} LC \rightarrow \pm RC \omega_{c} = 1 - \omega_{c}^{2} LC$$

$$LC \omega_c^2 \pm RC \omega_c - 1 = 0$$

$$\omega_{c} = \mp \frac{Rc}{2Lc} \pm \frac{1}{2Lc} \sqrt{(Rc)^{2} + 4Lc} = \mp \frac{Rc}{2Lc} \pm \frac{1}{2Lc} \sqrt{(Rc)^{2} \left[1 + \frac{4Lc}{(Rc)^{2}}\right]}$$

$$= \mp \frac{Rc}{2Lc} \pm \frac{Rc}{2Lc} \sqrt{1 + \frac{4L}{R^{2}c}} = \frac{R}{2L} \left[ \pm 1 \pm \sqrt{1 + \frac{4L}{R^{2}c}} \right] = \omega_{c}$$

$$\pm 4 \text{ Values: 2 are negatives}$$

of the other 2

12. [0, 4 points] What ideal filter type most closely approximates the frequency response of the circuit?

The 
$$|H(\omega)|^2$$
 curve rises  
on either side of  $\omega_{min} = \frac{1}{\sqrt{Lc}}$